

# Optimization of Biphase Codes for Pulse Compression Radars

B. Archana<sup>1</sup>, B. Shiny Sucharitha<sup>2</sup>, K. Amulya<sup>3</sup>, B. Nagaveni<sup>4</sup>, Chandrika Saxena<sup>5</sup>

<sup>1</sup>Department of Electronics & Communication Engineering, MGIT College, Gandipet, Hyderabad. E-mail: archies456@gmail.com

<sup>2</sup>Assistant Professor, Department of Electronics & Communication Engineering, St.Martin's Engineering College Hyderabad, Telangana, India. E-mail: shinylucky5@gmail.com

<sup>3</sup>Assistant Professor, Department of Electronics & Communication Engineering, Ashoka Institute of Engineering and Technology, Telangana, India. E-mail: amulukariki@gmail.com

<sup>4</sup>Assistant Professor, Department of Electronics & Communication Engineering, Ashoka Institute of Engineering and Technology, Telangana, India. E-mail: bnagaveni@gmail.com

<sup>5</sup>Assistant Professor, Department of Electronics & Communication Engineering, Marri Laxman Reddy Institution of Technology & Management, Telangana, India.  
E-mail: chandrika.s7march@gmail.com

**Abstract:** Radar designers utilize long pulse with a specific goal to acquire high energy for detection of the targets at long ranges and resolution of short Pulse by modulation of the long Pulse is known as Pulse compression. Through frequency or phase coding of the large bandwidth of signal can be achieved. In this technique, the performance of radar for a given application demands good autocorrelation property – to maintain low sidelobe levels at the matched filter output. One of the essential kinds of coding is the biphase code. Because these codes are easy to implement. The condition for the goodness of the transmitted codes depends upon the peak sidelobe level at the output of the matched filter. In this project, we tend to describe the optimization of biphase codes of larger length so that a high compression ratio can be achieved. Though the barker codes are the well-known biphase codes these are available only for lengths upto 13. It means with the help of barker code one can achieve the compression ratio only up to 13. Therefore, there is a need of searching higher length codes. In this work particle swarm optimization algorithm is used for the optimization of biphase codes up to 105. The results obtained are encouraging. For the code lengths, 60 to 64 the PSL = 4 and as length increases the results acquired are better. Further, the sidelobes are suppressed by designing a mismatched filter. It is shown that the sidelobes are appreciably at the cost of some signal-to-noise ratio.

**Keywords:** Particle Swarm Optimization (PSO) Algorithm, Biphase Codes, Autocorrelation.

## 1. INTRODUCTION

Sequences with peaky autocorrelation are widely used for RADAR applications. Achieving such sequences is a combination of non-comparable issues. So designing a signal having the above properties is a challenging problem. Many algorithms which are known as global optimization algorithms such as simulated annealing, tunneling algorithm, genetic algorithm, and particle swarm optimization algorithm, etc., only PSO was described in the literature. This project mainly focused on the designing of an optimum set of Biphase codes, which are optimized by using the Particle Swarm Optimization Algorithm.

Binary sequences having good autocorrelation that is low sidelobe levels are very useful in various applications such as in radar,[1] communication systems, and theoretical physics, etc.

The basic aim of using such sequences in radar systems is to achieve high range resolution. In fact, the resolution is stated as the ability of the radar to distinct two nearly spaced targets. When radar transmits such sequences, on receive, the output of the matched filter gives the compressed output that decides the resolution of the radar. The cost that we have to pay is sidelobes at the output of the matched filter. These sidelobes may mask the detection of weak target-present near the strong target. Therefore, the arrangement of the sequences is to be chosen in such a way that the peak sidelobes levels at the output of the matched filter must be minimum. Sets of codes that achieve the minimum peak sidelobes are of immense importance in radar applications. This project is an attempt to optimize the sequences having good autocorrelation properties using an optimization technique.

## 2. LITERATURE SURVEY

Biphase codes of lengthier series having low PSL and high Merit Factor (MF) are vital study area in the field of radar signal processing. The effectiveness of code design using PSO converges to autocorrelation shows the importance of the project. To attain decent range resolution we include PSO optimization that gives a decent autocorrelation of biphase codes in pulse compression radars.

Electronic counter measure (ECM) is the main threat in modern days. Anti-radiation missiles (ARM) are mostly endangering most of the military radars. Therefore, the Biphase coded signal having code agility can accomplish a low probability of intercept (LPI).

## 3. METHODOLOGY

In this project optimizes biphase code for pulse compression radar having low peak sidelobes (PSL). The toughness of the genetic algorithm degrades when code length increases. So, it is an attempt, to adopt the PSO algorithm for the generation of sequences of higher lengths. The PSO algorithm is founded based on biologically inspired and driven by social comparison. However, it is related to the Evolutionary Computation (EC) techniques, still, there are major modifications with genetic algorithms.

The PSO algorithm gives potential solutions to develop an approach to discovering an appropriate solution for a given problem. Being an optimization technique, the objective is to discover the global optimum solution of a real-valued function described in an above search space. In 1995 James Kennedy [9] and Russel Ehbart introduced the concept of Particle Swarm Optimization (PSO). However, it is inspired by the observed behavior of creatures in their natural habitations such as bird clustering or fish schooling. It means ultimately its origin is nature itself. The origins in natural processes, which lead the swarms, categorize the algorithm as Swarm Cleverness and Artificial Life.

### PSO Algorithm

Generally, the function that is needed to be minimized is  $f: \mathbb{R}^n \rightarrow \mathbb{R}$ . To this, function a candidate solution that is in terms of the vector is given as the input argument and the produced real number is taken as the output that indicates the candidate solution's objective function value [4]. The  $\nabla.f$  is unknown. The intention is to determine a solution such that  $f(\mathbf{a}) \leq f(\mathbf{b})$  for all  $\mathbf{b}$  in the search space, which is  $\mathbf{a}$  denoted by global minimum. By considering  $\mathbf{h} = -\mathbf{f}$  a maximum value can attain.

Here  $S$  denotes the number of particles in the swarm, which held the position  $\mathbf{x}_i \in \mathbb{R}^n$  in the search space and velocity  $\mathbf{v}_i \in \mathbb{R}^n$ . If  $\mathbf{p}_i$  is considered a well-recognized position of particle  $i$  and  $\mathbf{g}$  be the better-recognized position of the entire swarm.

The algorithm is given by:

Each particle  $i = 1, 2, 3, \dots, S$  does:  
 By initialization of particle's position with the evenly distributed random vector is given by  $\mathbf{x}_i \sim U(\mathbf{b}_{lo}, \mathbf{b}_{up})$ ,  $\mathbf{b}_{lo}$ , and  $\mathbf{b}_{up}$  are the lower and upper limits of the search space.  
 Particle's good recognized position is initialized [5] by its primary position:  $\mathbf{p}_i \leftarrow \mathbf{x}_i$   
 Swarm's good recognized position is updated if  $(f(\mathbf{p}_i) < f(\mathbf{g}))$ :  $\mathbf{g} \leftarrow \mathbf{x}_i$   
 Particle's velocity is primed:  $\mathbf{v}_i \sim U(-|\mathbf{b}_{up}-\mathbf{b}_{lo}|, |\mathbf{b}_{up}-\mathbf{b}_{lo}|)$   
 Till the termination condition is met (e.g. execute  $s$  number of iterations otherwise the solution with satisfactory above function value is established), recurrence:  
 1. Every particle  $i = 1, 2, 3, \dots, S$  do:  
 2. Each dimension  $d = 1, 2, 3, \dots, n$  do:  
 3. Random numbers are picked:  $r_p, r_g \sim U(0, 1)$   
 4. Particle's velocity to be updated:  $\mathbf{v}_{i,d} \leftarrow \omega \mathbf{v}_{i,d} + \phi_p r_p (\mathbf{p}_{i,d} - \mathbf{x}_{i,d}) + \phi_g r_g (\mathbf{g}_d - \mathbf{x}_{i,d})$   
 5. Particle's position is to be updated:  $\mathbf{x}_i \leftarrow \mathbf{x}_i + \mathbf{v}_i$   
 If  $(f(\mathbf{x}_i) < f(\mathbf{p}_i))$  do:  
 Particle's good recognized position is updated:  $\mathbf{p}_i \leftarrow \mathbf{x}_i$   
 If  $(f(\mathbf{p}_i) < f(\mathbf{g}))$  swarm's good recognized position is updated:  $\mathbf{g} \leftarrow \mathbf{p}_i$   
 6. The best solution found in  $\mathbf{g}$ .  
 The practitioner controls the behavior and efficiency of the PSO method selecting the factors  $\omega$ ,  $\phi_p$ , and  $\phi_g$ .

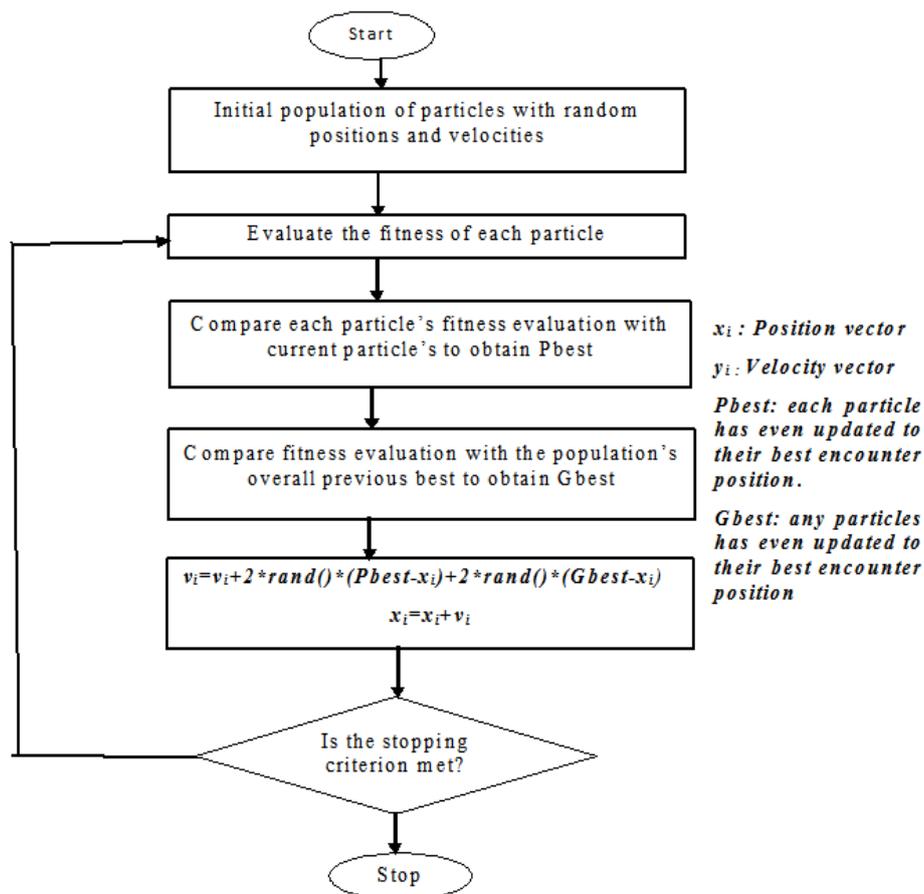


Figure 1: Working of PSO Algorithm

### Nested Barker Codes

For radar application, a lengthier sequence is preferred in order to attain a high pulse compression ratio. The longer code is obtained by combining two barker codes by Kronecker

product [2]. It can also be called a compound barker code. For example a 35-bit compound biphasic code is obtained by nesting 5-bit with 7-bit codes and the obtained 35 bit code is [1 1 1 -1 -1 1 1 -1 1 1 1 -1 -1 1 1 -1 1 1 -1 -1 -1 -1 1 1 -1 1 1 1 1 -1 -1 1 -1]. The peak sidelobes are not relatively diminished in spite of the greater compression ratio.

The only bi-phase code, which is having littlest sidelobe, is Barker code. The only odd length code 13 is the longest one which is a big disadvantage. The 13 length Barker code that attains mainlobe to sidelobe ratio of 22.8 dB that is smaller than 30dB practical value. In order to improve the ratio mismatched filter is proposed. The Nested code has a high PSR compared to barker codes. A nested binary code acquired by utilizing Kronecker product and its matched filter response are better. In the event that an N-bit Barker code is meant by BN, and another BM, then an MN bit code can be developed as  $BN \otimes BM$ . The Kronecker item is basically the BM code rehased N times, with every redundancy increased by the relating component of the BN code. For instance, a 65-bit code can be built as the item  $B13 \otimes B5$ . These codes have a peak sidelobe more prominent than 1.

#### Auto Correlation Property of Biphasic Codes

To obtain good autocorrelation property for the biphasic codes is optimized with the PSO algorithm. As per the PSO strategy initially, the best position and global best are assigned and modifies the velocity equation. The particle's motion is according to the velocity and the best sequence is extracted finally. The following equation determines the equations under the PSO algorithm:

$$x_{i,d}(it+1) = x_{i,d}(it) + v_{i,d}(it+1) \quad (1)$$

$$v_{i,d}(it+1) = v_{i,d}(it) + C_1 * Rnd(0,1) * [pb_{i,d}(it) - x_{i,d}(it)] + C_2 * Rnd(0,1) * [(gb_{i,d}(it) - x_{i,d}(it))] \quad (2)$$

Caption:

i particle's index, used as a particle identifier;

d dimension being considered, each particle[6] has a position and a velocity for each dimension;

It iteration number, the algorithm is iterative;

$x_{i,d}$  position of particle i in dimension d;

$v_{i,d}$  position of particle i in dimension d;

$C_1$  acceleration constant for the cognitive component;

Rnd stochastic component of the algorithm, a random value between 0 and 1;

$pb_{i,d}$  the location in dimension d with the best fitness of all the visited locations in that dimension of the particle i;

$C_2$  acceleration constant for the social component;

The paper "The particle swarm optimization algorithm" suggests parameters. Ioan Cristian Trelea does convergence study and parameter choice. And the parameters chosen are  $C_1 = 1.494$ ,  $C_2 = 1.494$  and random value chosen as 0.5.

When the velocity component utilized otherwise atleast multiplied with factor W though it cannot simply modify its velocity concerning the best solution but tends to discover new fangled regions of search space. It enables the discovery of new-fangled spaces according to time "spend counteracting" the former momentum when it first "counteracts" the gained momentum previously. By multiplying, the previous velocity to weight factor the above variation achieved. To acquire a better ratio of performance progress and the successful algorithm's in discovering the desired solution W value chosen between [0.9, 1.2].

#### Mismatched Filter

Further to reduce the sidelobes of the optimized sequences, a mismatched filter is designed in this section. The reduction or suppression of sidelobe is achieved by sacrificing some amount of signal-to-noise ratio (SNR). Let the biphasic sequence is given by

$$A = \{a_1, a_2, a_3, \dots, a_N\} \quad (1)$$

The filter elements are

$$H = \{h_1, h_2, h_3 \dots \dots h_M\} \quad (2)$$

Where the elements are real and the relationship is maintained as  $N \leq M$ . In addition to this condition, another condition is that if N is odd then M must be odd, and if we consider N even, M also must be even. This specifies that (M- N) is always even, and  $(M-N)/2 = Z$ , which is an integer. We are defining that the Z is an all-zero sequence of length n. Now a zero-padded signal sequence can be created whose length is  $(M = N + 2n)$ . The final sequence is given by

$$S = \{Z A Z\} \quad (3)$$

The above equation gives an explanation for a simple mismatch filter. By using the windowing function even sidelobes can be further suppressed. This is clearly observed from the fig3 the sidelobe suppression increases as the length of 'S' increases. Length of the sequence results H and S are equal and that is equal to M. It is also kept in mind the filter is designed in such a way that the cross-correlation peak appears at zero delay ( $\tau = 0$ ). The cross-correlated output is shown in fig.1 can observe that the sidelobes are suppressed at the cost of SNR loss. Mismatch loss for different filter M is given in Table 1.

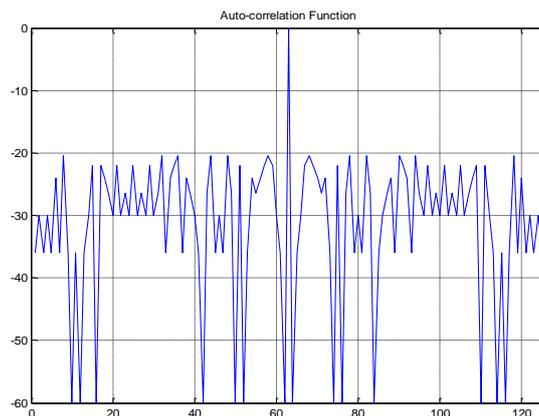


Figure 2 ACF of N=63, PSO (Matched Filter)

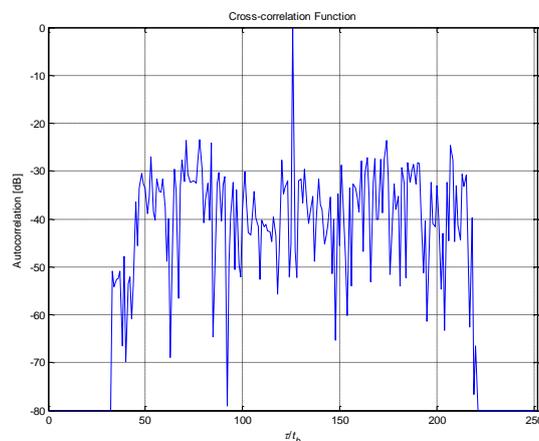


Figure 3 ACF of N=63, PSO (Mismatched Filter length M=126) mismatch loss=1.02db

Table 1: Results obtained for length 63,126,189,252

|                           |     |      |      |      |
|---------------------------|-----|------|------|------|
| Filter Length (M)         | 63  | 126  | 189  | 252  |
| SNR Loss (dB)             | 0   | 1.02 | 1.34 | 1.83 |
| Sidelobe Suppression (dB) | -20 | -24  | -30  | -35  |

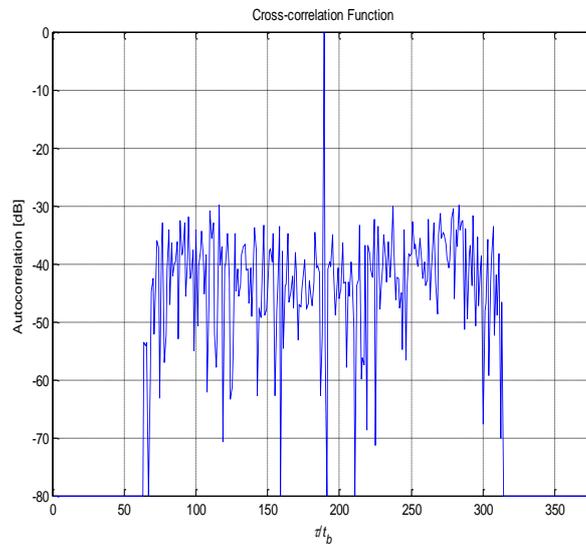


Figure 4 ACF of N=63, PSO (Mismatched Filter length M= 189) mismatch loss = 1.34db

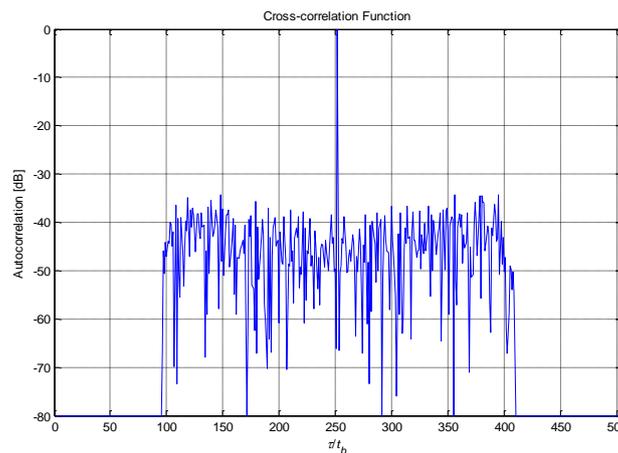


Figure 5 ACF of N=63, PSO (Mismatched Filter length M= 252) mismatch loss = 1.83db

#### 4. RESULTS AND DISCUSSION

Computational difficulty in generating codes and its restriction on code length can be negated by using an optimization algorithm. In this project Particle swarm optimization algorithm [4] is used to generate codes with length up to 105. On observation, we can show how the sidelobe levels are reduced with the best correlation property.

The following project is related to the particle swarm optimization the following are the results observed for different lengths with its phases.

The results are noticed for various lengths such as  $n=61,62,63,64,65,66,66,67,-----101,102,103,104,105$ . And its PSL value is also observed for certain lengths. For this matlab R2015a is utilized.

For sig-len=31  
 mainlobe=31

sidelobe=2.0000  
 df=15.5000  
 x=[1 -1 -1 1 -1 -1 1 -1 -1 1 1 1 -1 -1 -1 -1 -1 1 1  
 1 -1 1 1 1 1 1 -1 1 -1 1]

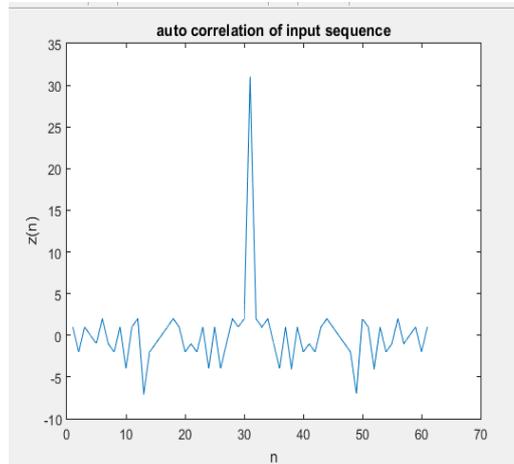


Figure 6 ACF of input length sequence 31

- sig-len=60  
 mainlobe=60  
 sidelobe=4.0000  
 df=15.0000  
 x=[-1 1 1 1 -1 1 1 1 -1 -1 1 1 -1 1 1 -1 1 -1 -1 -1  
 1 1 -1 1 1 1 1 1 -1 1 -1 -1 1 -1 1 1 1 -1 1 -1 1  
 1 -1 -1 -1 -1 -1 -1 1 -1 -1 -1 1 1 -1 1 -1 1 1 1]

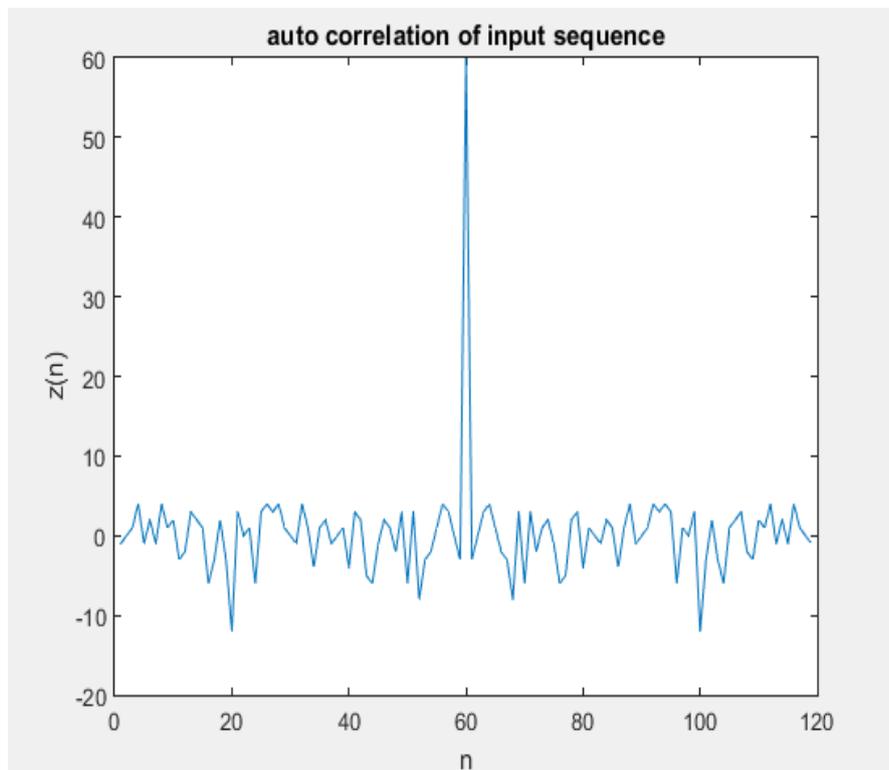


Figure 7 ACF of input length sequence 60

The results for a length up to 105 are recorded and the sequence is a combination of 1 and -1 but it is rearranged as 1 and 0 respectively and sequence are represented in hexadecimal.

Table 2: Results obtained up to length 105

| Length of sequence | Sidelobe | PSL          | ISL          | Merit factor | Sequence              |
|--------------------|----------|--------------|--------------|--------------|-----------------------|
| 61                 | 5        | -<br>21.7271 | -<br>10.1921 | 10.4522      | E653A2FE4907943       |
| 62                 | 5        | -<br>21.8684 | -<br>10.6104 | 11.5090      | 11AE87BB2701699C      |
| 63                 | 4        | -<br>23.9456 | -<br>11.5464 | 14.2770      | 6583F32E3BB48422      |
| 64                 | 5        | -<br>22.1441 | -<br>10.8088 | 12.0471      | 42173660D574B7B8      |
| 65                 | 4        | -<br>24.2170 | -<br>10.6952 | 11.7361      | 27E486B85858E6DA      |
| 66                 | 5        | -<br>22.4114 | -<br>10.8038 | 12.0331      | 3112CB76BDB831DC1     |
| 67                 | 5        | -<br>22.5420 | -<br>10.3515 | 10.8430      | 354D9F08F8AC25BEB     |
| 68                 | 5        | -<br>22.6707 | -<br>10.2953 | 10.7037      | 9B02AE47ABCA19259     |
| 69                 | 5        | -<br>22.7975 | -<br>10.1118 | 10.2608      | C5C9F0B2D9FA439E8     |
| 70                 | 5        | -<br>22.9225 | -<br>10.5271 | 11.2903      | 10B9A03D6791966EB5    |
| 71                 | 5        | -<br>23.0457 | -<br>10.6503 | 11.6152      | C29B63D5AA980FBB8     |
| 72                 | 5        | -<br>23.1672 | -<br>10.0400 | 12.7059      | 9543D7C80A41B631C9    |
| 73                 | 5        | -<br>23.2870 | -<br>10.7537 | 11.8951      | 10357397D3B606E9AD4   |
| 74                 | 5        | -<br>23.4052 | -<br>11.2568 | 13.3561      | A191D7F7A47B85DA6C    |
| 75                 | 5        | -<br>23.5218 | -<br>10.8926 | 12.2817      | 4BC6A87CDBD022E309E   |
| 76                 | 5        | -<br>23.6368 | -<br>10.9511 | 12.4483      | 11EB5E8B2270C0E414F   |
| 77                 | 5        | -<br>23.7504 | -<br>11.1022 | 12.8891      | 193EE09BF0B02953DD4E  |
| 78                 | 5        | -<br>23.8624 | -<br>10.9755 | 12.5185      | 3D233FA9B55E606116D7  |
| 79                 | 5        | -<br>23.9731 | -<br>11.2687 | 13.3927      | 1F9970BCD12EEC25099E  |
| 80                 | 6        | -<br>22.4987 | -<br>11.1776 | 13.1148      | 1E22CC21C906B7D9958E8 |
| 81                 | 5        | -            | -            | 13.5558      | 4CAF34C443AFB94B82D0  |

|     |   |              |              |         |                             |
|-----|---|--------------|--------------|---------|-----------------------------|
|     |   | 24.1903      | 11.3212      |         |                             |
| 82  | 5 | -<br>24.2968 | -<br>11.4820 | 14.0669 | 17274F41DE2CBBDC44825       |
| 83  | 6 | -<br>22.8185 | -<br>11.2383 | 13.2992 | D3BD3FA09AAED13A321C        |
| 84  | 6 | -<br>22.9225 | -<br>11.2265 | 13.2632 | 11182BDC2C37EDD9D6585       |
| 85  | 6 | -<br>23.0253 | -<br>10.7493 | 11.8832 | 1E92A206D2AB778D277C72      |
| 86  | 6 | -<br>23.1269 | -<br>11.3180 | 13.5458 | 958110AFB84F363D91AA7       |
| 87  | 6 | -<br>23.2273 | -<br>10.8805 | 12.2476 | 11CCC4C951A7A4BFAC550F      |
| 88  | 5 | -<br>24.9102 | -<br>11.1372 | 12.9933 | 88D52B6FD618D17C407926      |
| 89  | 6 | -<br>23.4247 | -<br>11.5684 | 14.3496 | 1E6396E1F64E5D58009AD46     |
| 90  | 5 | -<br>25.1054 | -<br>11.4656 | 14.0138 | 2BCB1CB13FB25C0316C453F     |
| 91  | 6 | -<br>23.6178 | -<br>11.0785 | 12.8189 | 784F595DBB5307BE9822503     |
| 92  | 6 | -<br>23.7127 | -<br>11.4082 | 13.8301 | 557D4B9B4D6A746618606F      |
| 93  | 6 | -<br>23.8066 | -<br>11.0191 | 12.6447 | 1B97E2A24B064C7A866157D0    |
| 94  | 6 | -<br>23.8995 | -<br>10.9992 | 12.5869 | 1A973D38AE024EE594377496    |
| 95  | 6 | -<br>23.9914 | -<br>11.4521 | 13.9706 | 1E66A5CBDECE8A2487F40BB5    |
| 96  | 6 | -<br>24.0823 | -<br>11.0241 | 12.6593 | E646CE11035D426BE47E52E5    |
| 97  | 6 | -<br>24.1724 | -<br>10.9045 | 12.3154 | 19848D81169E65BE763A8BEE4   |
| 98  | 6 | -<br>24.2614 | -<br>11.2875 | 13.4510 | 17242BDDDF383533A17839E8A6  |
| 99  | 6 | -<br>24.3496 | -<br>11.6261 | 14.5415 | 7E1340620D4E1AB69AB9E8947   |
| 100 | 7 | -<br>23.0980 | -<br>11.3312 | 13.5870 | F2405D453F4CEA7B6654A86C1   |
| 101 | 7 | -<br>23.1844 | -<br>10.9695 | 12.5012 | 124E73CCA2F66AD89028E0CF64  |
| 102 | 7 | -<br>23.2700 | -<br>11.8854 | 15.4362 | A2E5DCBE6728028DEC0DAC1AF   |
| 103 | 7 | -<br>23.3547 | -<br>11.5294 | 14.2212 | 9A6B395AC5E8BB64484C7847    |
| 104 | 7 | -<br>23.4387 | -<br>11.0567 | 12.7547 | 4D0FA54AFFB90764E628E06BA4  |
| 105 | 7 | -<br>23.5218 | -<br>11.3069 | 13.5110 | 1E9D6C21A2A6F79B159F6393762 |

## 5. CONCLUSION

The effort of extracting the best frequency order is increasing with the increase in the length of code. In order to lessen the time we took an approach for optimized search instead of an exhaustive search. The decent autocorrelation property is achieved by a particle swarm optimization algorithm. The main objective of the project is to develop decent codes with better correlation properties. That is why we have chosen autocorrelation property. This can be prolonged to the confrontation between autocorrelation and cross-correlation. Further can be extended for a longer length and better results are achieved by using a mismatched filter. Many other optimization techniques such as GA, Stimulated Annealing can be used.

## 6. REFERENCES

- [1] <https://dattasainathd.github.io/>
- [2] <https://www.intechopen.com/books/swarm-intelligence-recent-advances-new-perspectives-and-applications/particle-swarm-optimization-a-powerful-technique-for-solving-engineering-problems>.
- [3] Mahafza, Bassem R.-Radar Systems Analysis, and Design using Mat lab (3<sup>rd</sup> Edition)- Taylor and Francis-2013.
- [4] Zhang, Y. (2015). "A Comprehensive Survey on Particle Swarm Optimization Algorithm and Its Applications". *Mathematical Problems in Engineering*. 2015: 931256.
- [5] Taherkhani, M.; Safabakhsh, R. (2016). "A novel stability-based adaptive inertia weight for particle swarm optimization". *Applied Soft Computing*. 38: 281–295. DOI:10.1016/j.asoc.2015.10.004.
- [6] R.C. Eberhart and Y. Shi. Uses and properties: Particle swarm optimization: In Proceedings of the 2001 congress on evolutionary computation, volume 1 pages 81-86. Piscataway, NJ, USA: IEEE, 2001
- [7] Irina Gladkova, "Families of Waveforms with good auto and cross Ambiguity properties", *IEEE Transactions on Aerospace and Electronic Systems* 0-7803- 6707-3/01/\$10.00 Q 2001 IEEE.
- [8] Skolnik, M.I., "Introduction to Radar Systems (3rd ed.)." New York: McGraw-Hill, 2001.
- [9] <https://www.researchgate.net/publication>

## BIOGRAPHIES



**B. Archana** received the Bachelor of Technology degree in Electronics & Communication Engineering from ICFAI University for higher education, IFHE, 2014, and Masters of Technology degree in Digital Electronics and Communication Systems from Mahatma Gandhi Institute of Technology (Jawaharlal Nehru Technological University Hyderabad) in 2019. My area of interest is in Digital Signal Processing, VLSI Design.



**B. Shiny Sucharitha** is working as an Assistant Professor in the department of ECE, St.Martin's Engineering College, Secunderabad, India. Received the Bachelor of Technology degree in Electronics & Communication Engineering from DRK Institute of Science and Technology (Jawaharlal Nehru Technological University Hyderabad) in

2016 and a Master of Technology degree in Digital Electronics and Communication System from Mahatma Gandhi Institute of Technology (Jawaharlal Nehru Technological University Hyderabad) in 2019. My area of interest is in Digital Signal Processing, Communication. She has published papers in International Journals/Conferences.



**Karika Amulya** is working as an Assistant Professor in the department of ECE, Ashoka Institute of Engineering and Technology, Hyderabad, India. Received Bachelor of Technology in Electronics and Communication from BVRIT College of Engineering for Women, Hyderabad (Jawaharlal Nehru Technological University Hyderabad) in 2016, and Masters of Technology degree in Digital Electronics and Communication System from Mahatma Gandhi Institute of Technology (Jawaharlal Nehru Institute of Technology, Hyderabad) in 2019. Area of Interest is in Image Processing, Communications and Digital Signal Processing.



**B. Nagaveni** is working as an Assistant Professor in the department of ECE, Ashoka Institute of Engineering and Technology, Hyderabad, India. Received Bachelor of Technology in Electronics and Communication from Gates Institute of Technology, Gooty, (Jawaharlal Nehru Technological University Anantapur) in 2010, and Masters of Technology degree in VLSI Design from Srinivasa Ramanujan Institute of Technology (Jawaharlal Nehru Institute of Technology, Anantapur) in 2013. Area of Interest is in VLSI design, Digital Communications.



**Chandrika Saxena** is working as an Assistant Professor in the department of ECE, Marri Laxman Reddy Institution of Technology and Management, Hyderabad. Received the BE from Rajiv Gandhi Proudhyogiki Vishwavidyalaya, Madhya Pradesh, India, and M.tech from Rajiv Gandhi Proudhyogiki Vishwavidyalaya in 2015, Madhya Pradesh, India. The area of interest is in Digital Signal Processing.